

Antenna Performance Degradation Resulting From Aviation Fuel Exposure

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Abstract

Often, antennas operate within a hostile environment that potentially alters the antenna's electrical performance. A case study was conducted to evaluate the performance of an antenna in the presence of aviation fuel such as JP-5 using HFSS (high frequency, finite element method, structure simulator.) An extensive research effort was conducted to obtain the JP-5 aviation fuel's electrical performance characteristics at microwave and millimeter frequencies.

Two Archimedean planar cavity-backed spiral antennas and associated microstrip and coaxial tapered baluns were designed for right hand polarization (RHCP). A quartz fiber radome was also designed and included in the model.

Formulation for a dielectric composite mix material scenario was used to model cases where material and fuel mixed in various concentrations. Several mathematical models for the estimation of the effective dielectric constant and loss tangent resulting from composite materials were evaluated. These included but were not limited to the Finite Element, Maxwell Garnett, and Bruggeman methods.

The enclosed presentation is a brief summary presentation extracted from extensive research conducted in 2005.

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- **Dr. Dejan Filipovic** from University of Colorado, Boulder, Colorado and **Brad Brim** from Ansoft Corporation for providing common design examples of dual arm, planar spiral antennas.

From Raytheon Company, Space and Airborne Systems, EW, Goleta, California.

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- Finally, **William W. Wallace** from the Naval Air Systems Command, Patuxent River, Maryland, for providing information documents on JP-5 fuel.

DIELECTRIC COMPOSITE MATHEMATICAL METHODS

An excellent summary on the theories for the dielectric constant can be found in reference number 1. The study reviews dielectric constant mathematical derivations methods by Onsager, Kirkwood and Cole¹. After reviewing numerous papers available in literature, the effective dielectric material composite formulation used in this report was selected from those mathematical methods described in reference 2 and 3. These formulas are widely accepted in industry and are listed below:

Maxwell-Garnet formula

$$\epsilon_{av} = \epsilon_m \left[1 + \frac{3f \left(\frac{\epsilon - \epsilon_m}{\epsilon + 2\epsilon_m} \right)}{1 - f \left(\frac{\epsilon - \epsilon_m}{\epsilon + 2\epsilon_m} \right)} \right] \quad (1)$$

where ϵ_m = dielectric constant of matrix; ϵ = dielectric constant of inclusion; f = volumetric fraction of the inclusion.

The other formulation is the Bruggeman

$$f_1 \frac{\epsilon_1 - \epsilon_{av}}{\epsilon_1 + 2\epsilon_{av}} + (1 - f_1) \frac{\epsilon_2 - \epsilon_{av}}{\epsilon_2 + 2\epsilon_{av}} = 0 \quad (2)$$

Where, ϵ_1 and ϵ_2 are the dielectric constant of the composite materials and f_1 is the volume occupied by dielectric material of ϵ_1 . Note that Bruggeman formula is for symmetrical cases and the Maxwell Garnett formula is reported³ to be less accurate for volumetric fraction greater than 0.3.

The above formulas were reformulated and programmed in Excel. Other approaches to determine the composite materials electrical properties are by the use of finite element (HFSS,) and/or finite difference methods. Several publications on these techniques are available^{4,5}. These references also address the effective loss tangent.

Other approaches were also used in this analysis to determine the composite material electrical performance. These are briefly described below:

With the use of finite element method (HFSS) the following geometry was modeled in a small size sample:

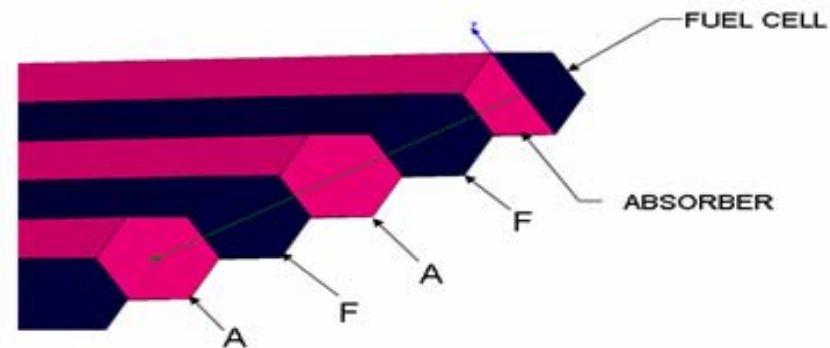


Figure 12

The cell size was chosen to be equal to $2\pi r/\lambda \approx 0.1$, where r is the longest length of the hexagonal face and λ is the wavelength. Smaller size hexagonal tubing was tried. When compared to available measured data, this approach appears to yield a surface effects rather than a volumetric material mix effect. There is more discussion that merits attention on the dielectric subject such as the Debye relation and the dielectric characteristic of a material⁶, but this is beyond the scope of this report. Additional studies about permittivity of material compounds were considered with improved results like the finite element method (FEM) and the Monte Carlo method⁷. For example, the previously cited reference indicated that the effective loss tangent ($\tan\delta$) and the effective permittivity (ϵ_e) to be logarithmic related as shown below:

$$\tan\delta = A \cdot \log(\epsilon_e) + B \quad (3)$$

where A and B are numerical constants used for curve fitting.

Finally, an additional formulation was developed by Tom Debski of Raytheon, EW in Goleta. This formulation is interesting because it seems to predict the effective dielectric constant of a composite material mix similar to the power-law and Bruggeman empirical models.

Debski formula:

$$\epsilon_{\text{effective}} \approx \left[f \sqrt{\epsilon} + (1-f) \sqrt{\epsilon_m} \right]^2 \quad (4)$$

Where f is the volumetric fraction, and ϵ and ϵ_m are the dielectric constants of the two materials.

Bruggeman Relationship Formulation for the Excel program

$$f \times ((E_1 - E_{av}) \div (E_1 + 2E_{av})) + (1 - f) \times ((E_2 - E_{av}) \div (E_2 + 2E_{av})) = 0$$

$$f \times ((E_1 - E_{av}) \times (E_2 + 2E_{av})) + (1 - f) \times ((E_2 - E_{av}) \times (E_1 + 2E_{av})) = 0$$

$$f \times [E_1 E_2 + (-E_2 + 2E_1) \times E_{av} - 2E_{av}^2] + (1 - f) [E_1 E_2 + (2E_2 - E_1) \times E_{av} - 2E_{av}^2] = 0$$

$$-2E_{av}^2 + 2[f \times (E_1 - (E_2 \div 2)) + (1 - f) \times (E_2 - (E_1 \div 2))] \times E_{av} + E_1 E_2 = 0$$

$$E_{av}^2 - [f \times (E_1 - (E_2 \div 2)) + (1 - f) \times (E_2 - (E_1 \div 2))] \times E_{av} - (E_1 E_2 \div 2) = 0$$

$$B = f \times (E_1 - (E_2 \div 2)) + (1 - f) \times (E_2 - (E_1 \div 2)), C = (E_1 E_2) \div 2$$

$$\text{Note: } x^2 - Bx - C = 0 \rightarrow x = (B \pm \sqrt{B^2 + 4C}) \div 2$$

$$\underline{E_{av} = (B + \sqrt{B^2 + 2E_1 E_2}) \div 2} \rightarrow \text{because...} E_{av} > 0$$

Dielectric Mixing Equations

		ϵ	ϵ_m	f				
		2	3	0.5				
first equation		$\epsilon_{av} = \epsilon_m * (1 + 3 * f * (\epsilon - \epsilon_m) / (\epsilon + 2 * \epsilon_m)) / (1 - f * (\epsilon - \epsilon_m) / (\epsilon + 2 * \epsilon_m))$						
Maxwell-Garnett								
		ϵ_{av}						
first equation solved		2.4706						
					FOR 50% MIX OF FUEL AND ABSORBER			
		ϵ_1	ϵ_2	f				
		2	3	0.5				
second equation		$0 = f * ((\epsilon - \epsilon_{av}) / (\epsilon + 2 * \epsilon_{av})) + (1 - f) * ((\epsilon_m - \epsilon_{av}) / (\epsilon_m + 2 * \epsilon_{av}))$						
Bruggeman								
		ϵ_{avg}	ϵ_{ee} (intermediate variable)					
second equation solved		2.4664		-1.25				
kerosene	Er	LT					DIEL MIX	LT

Material Research

- Extensive research was conducted to determine the composition and properties of JP-5 fuel.
- First, it was determined that JP-5 fuel is specially refined kerosene and is part of the hydrocarbon liquid family. Although JP-5 contains several additives like antioxidant, corrosion, and icing inhibitors, the author decided not to include the additives in the analysis. This was based on comparison of kerosene data versus JP-5 at some discrete frequencies.
- Second, the conductivity⁸ of JP-5 is in the range of 150 to 600 pS/m at room temperature. This is hardly a conductive substance. Nevertheless, the conductivity was included in the analysis.
- Measurements show that the dielectric constant of JP-5 fuel changes in a linearly versus temperature^{9,10}. At 400 Hz, the dielectric constant change from -60° to +260° was 2.25 to 2.08, compared to 2.15 at room temperature. This constitutes a percent change of 8%. Since the dielectric constant varies linearly with temperature, the dielectric change as a result of temperature appears to be independent of frequency. Therefore, the dielectric constant versus frequency of the JP-5 fuel for the HFSS model was varied $\pm 8\%$. The loss tangent was also changed but not to the same $\pm 8\%$ factor.
- The antennas performance parameters evaluated were the gain, beam squint, and beamwidth at microwave and millimeter frequencies. The electrical properties of JP-5 fuel found in literature^{11,12,13,14,15,16} for the final HFSS model were altered to represent to the worst dielectric variation versus frequencies.
- Pressure also impacts the dielectric constant value, but this factor was determined to have minimal influence, and was not considered in the current analysis.

Material Research

- The "Dielectric Materials and Applications" book by Arthur von Hippel is an excellent source for petroleum oils (JP-1, JP-3, Kerosene,) but only at 3 GHz. This source reported Kerosene (\approx JP-5) $\epsilon_r = 2.09$ and $\tan\delta = 0.0045$. Another source "Fuel Properties" NAVAIR report number 06-05-06 provided by Naval Air Systems Command, Patuxent River, Maryland, site reported a JP-5 dielectric $\epsilon_r = 2.05$ at 400 Hz. This source appear to agree with "Study to Determine the Electrical and Physical Properties of Aviation Fuel," by C.C. Petersen, WADC report 52-53, Wright Air Development Center. Another source¹⁷ reported an $\epsilon_r = 2.086$ at 9.538 GHz. An Interesting study on Unpolar liquid kerosene ϵ_r and $\tan\delta$ by C. Cotae and Gh. Calugaru¹⁸ (Department of Physics, Polytechnic Institute, Jassy, Romania) was also considered in the determination of Kerosene electrical properties. Additional JP-5 sources were evaluated but not referenced in this report. For high frequency¹⁹ (millimeter band as an example) data on electrical properties of fluids, the study conducted by V.V. Meriakri, "Low-loss Materials for Application in Millimetre and Submillimetre Wave Ranges, was used to gain insight to non-polar liquids electrical performance at mm wave frequencies.
- The loss tangent ($\tan\delta$) and dielectric constant (ϵ_r) values located in publications on JP-5 and Kerosene were altered based on other factors like temperature and in house formulation calculations. For example, the 2, 4, and 6 mil boundary impedance surface layer calculations for kerosene at microwave and millimeter frequency band revealed that the real part of the impedance hardly changed..

Spiral Antenna

Radome Design and Materials Used for the Analysis

- Two Archimedean dual arm spiral antennas for microwave and millimeter frequency bands were designed. The spirals were designed for right hand circular polarization only. The left hand circular polarization was not addressed since the fuel effects should be the same for both polarizations. The design consisted of a spiral, dual arm gold element, and a balun impedance transformer. These components were supported by honeycomb structure with free space foam under the elements. The design was based on common knowledge available in literature^{20,21,22,23,24,25,26}. Both spiral antennas were validated with HFSS (baseline data shown in page 16). Two quartz fiber radomes^{27,28} combined into one structure were also designed
- The antennas and associated components are depicted by Figures 1, 2 and 3. The honeycomb material used between the microwave spiral element and housing bottom plate was not carbon loaded. The honeycomb dielectric constant for the vertical wall was 1.042 and 1.06 for the horizontal geometry. The loss tangent was 0.00083 and 0.00036 respectively. The Rohacell HF 71H low loss foam ($\epsilon_r \approx 1.05$ at 2 GHz and 1.042 at 30 GHz with loss tangent ≈ 0.0002 to 0.016 at 30 GHz) was used between the honeycomb and element substrate. The substrate is an 8 mil thick Roger RT/duroid 5880. High loss absorber within spiral back cavity similar to FDS and GDS were also used within the spiral back cavity. The data for the frequency of interest can be obtained from Rogers Corporation. The radome electrical material properties and other materials were programmed in HFSS with frequency dependent material function. The HFSS models were solved at individual frequencies increments. The convergence parameters were maintained constant from the baseline model versus the fuel layers models.

TYPICAL MICROWAVE SPIRAL ANTENNA (Cut Plane View)

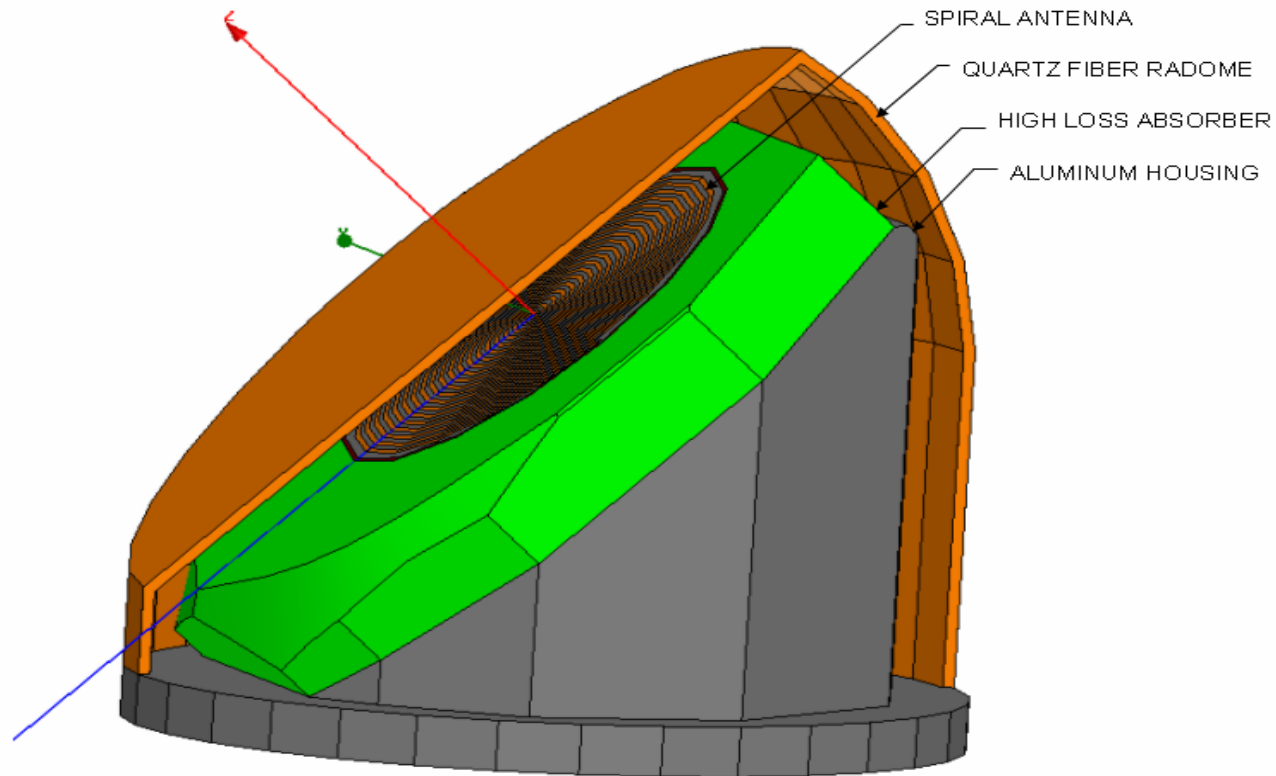


Figure 1

TYPICAL MICROWAVE SPIRAL ANTENNA

(View of internal components)

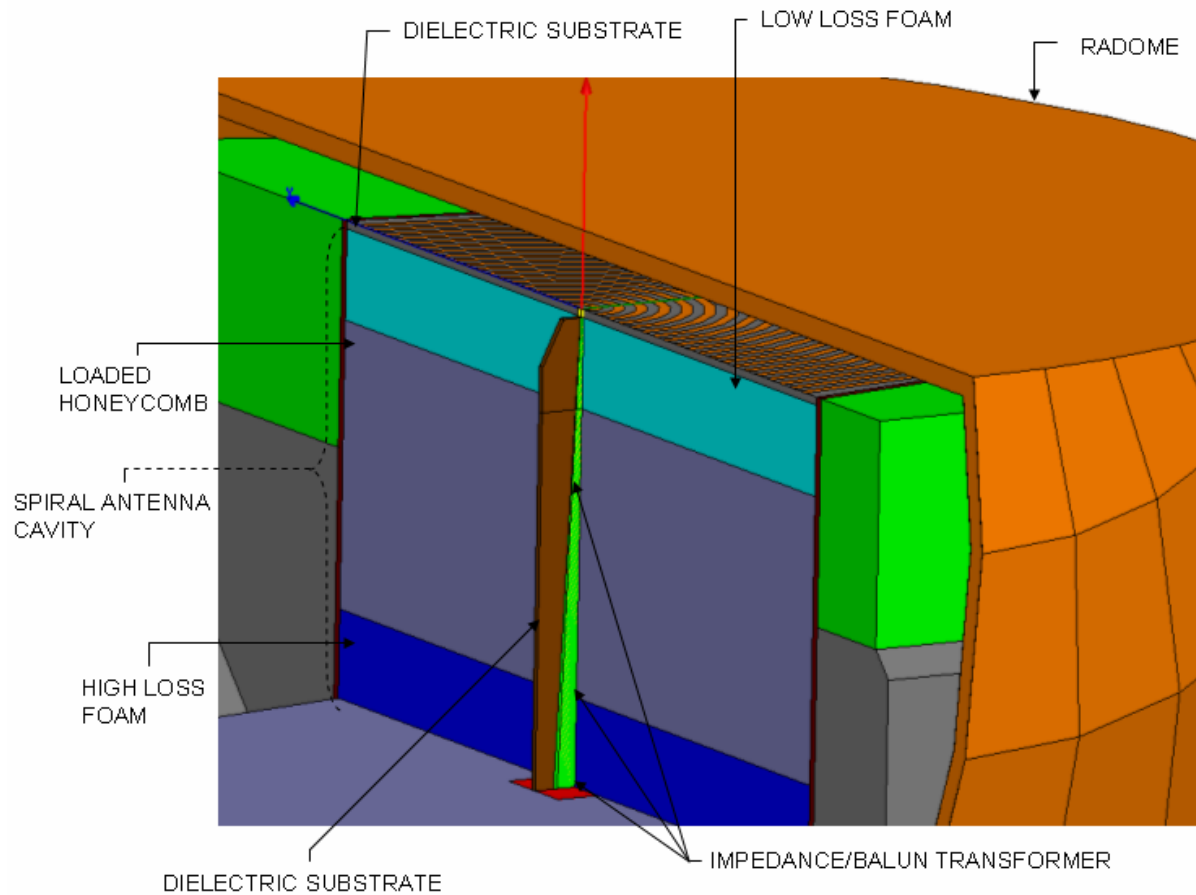
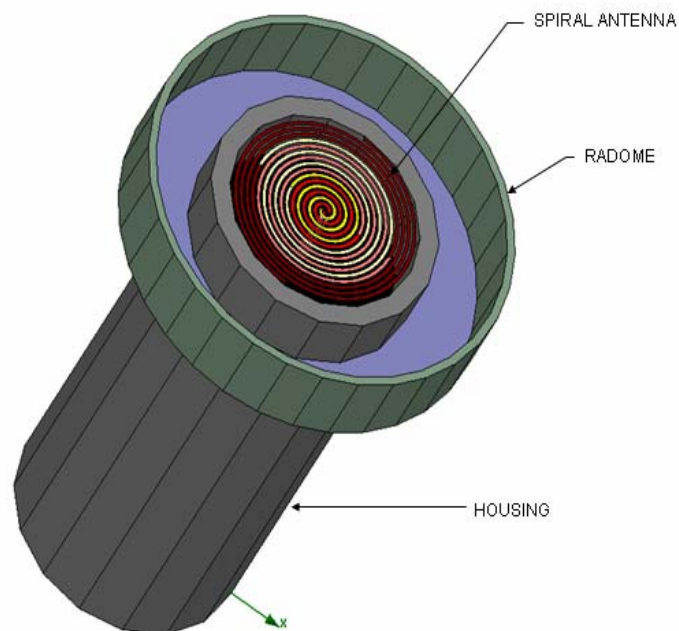
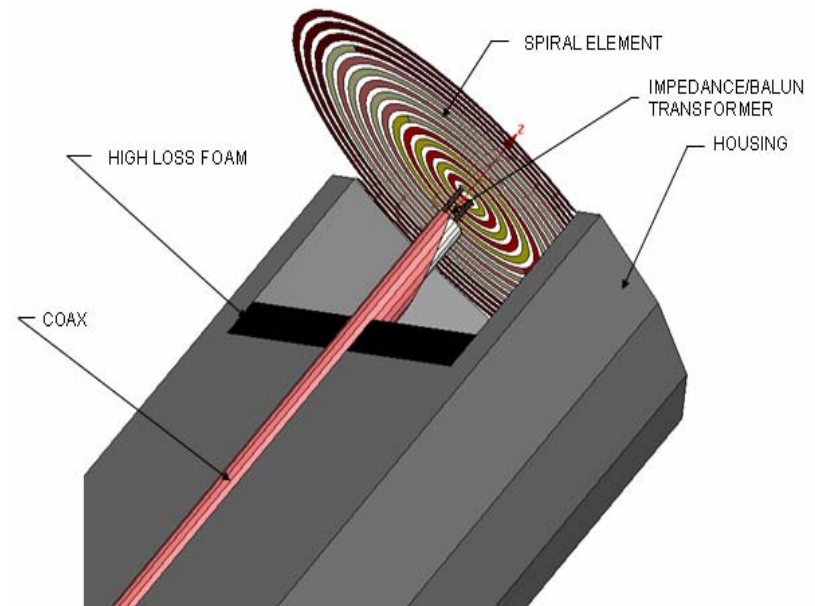


Figure 2

TYPICAL MILLIMETER SPIRAL ANTENNA



(a)



(b)

Figure 3

BASELINE 3D RADIATION PATTERNS

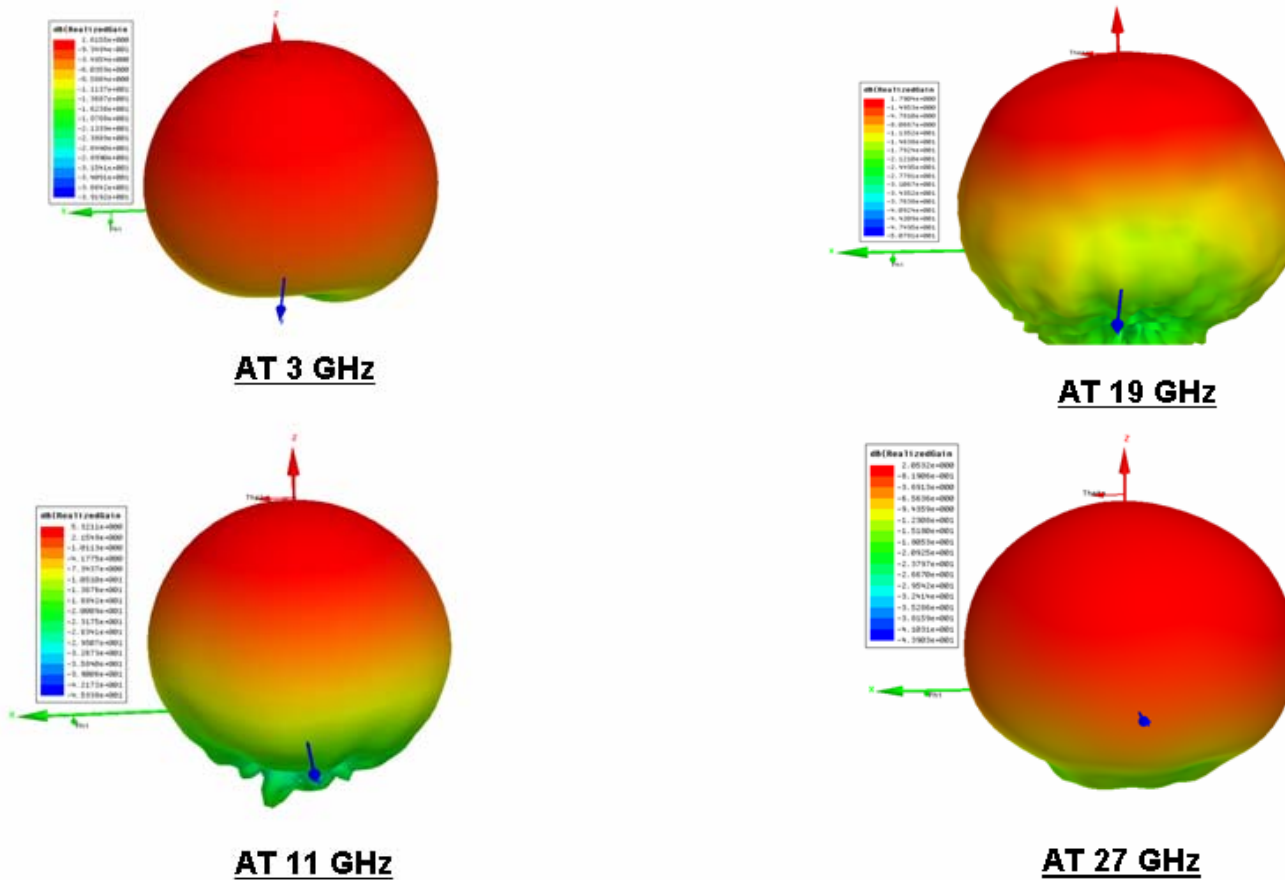


Figure 4

EXAMPLES OF ANALYSIS SCENARIO

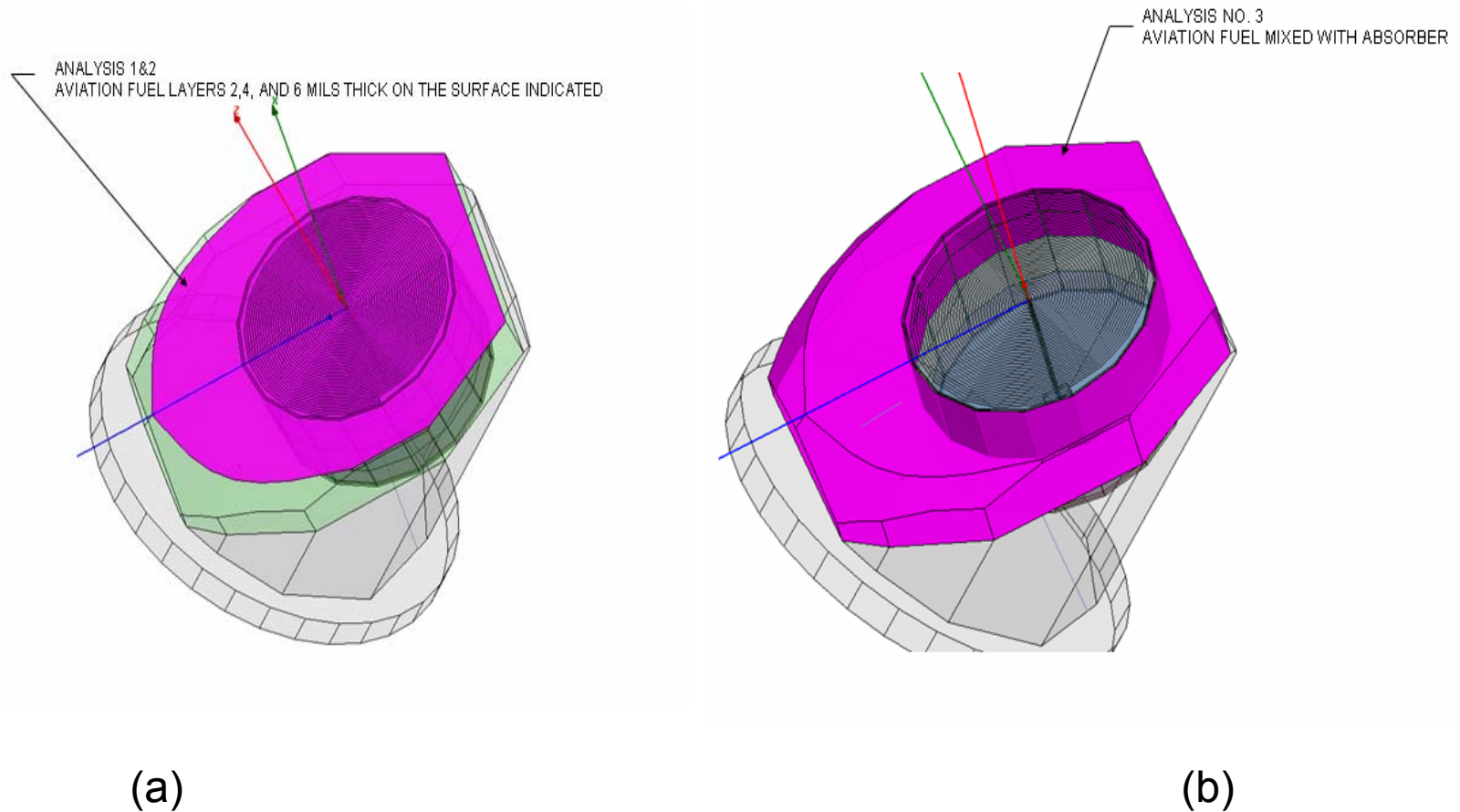


Figure 5

ANALYSIS RESULTS

KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. PEAK REALIZED GAIN AT MICROWAVE FREQUENCIES

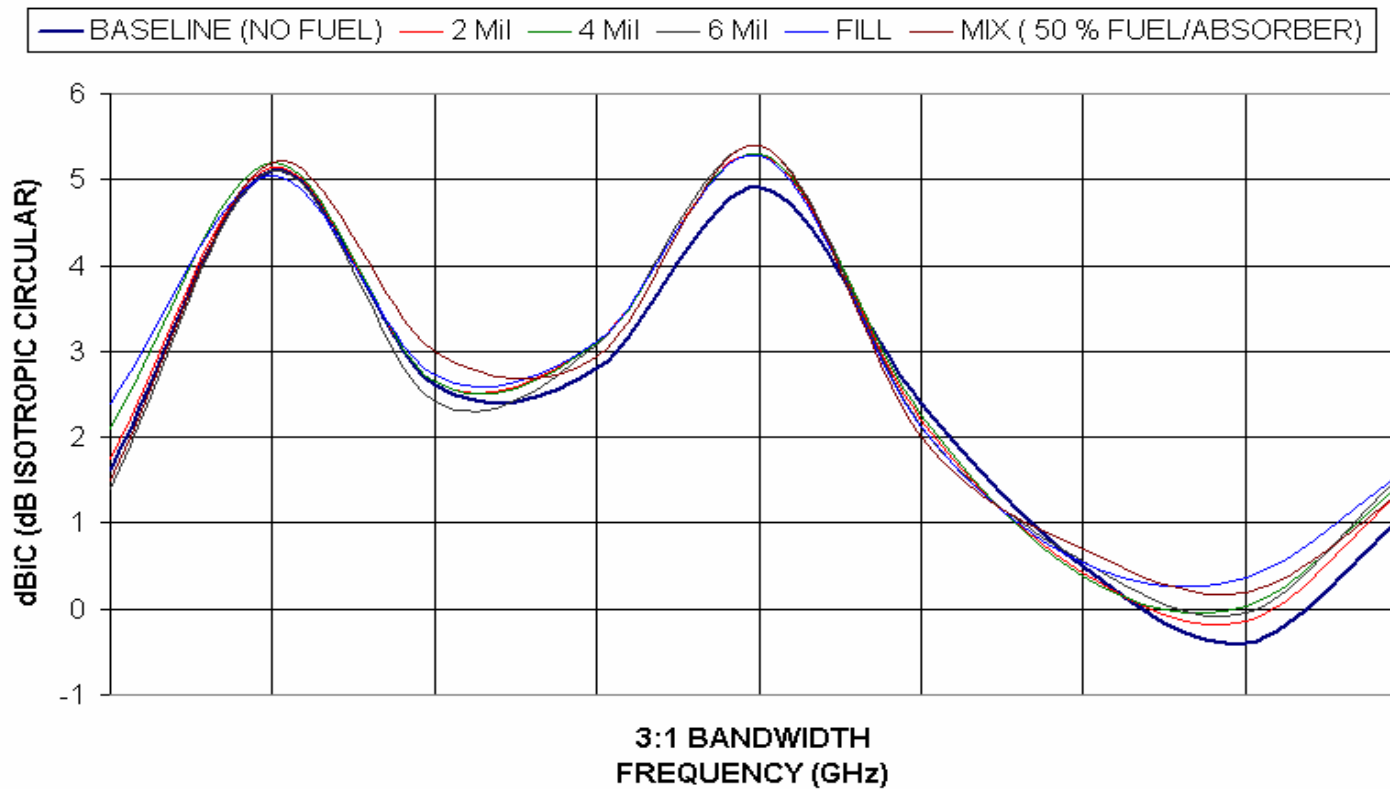


Figure 6

ANALYSIS RESULTS

KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. PEAK REALIZED GAIN AT MILLIMETER FREQUENCIES

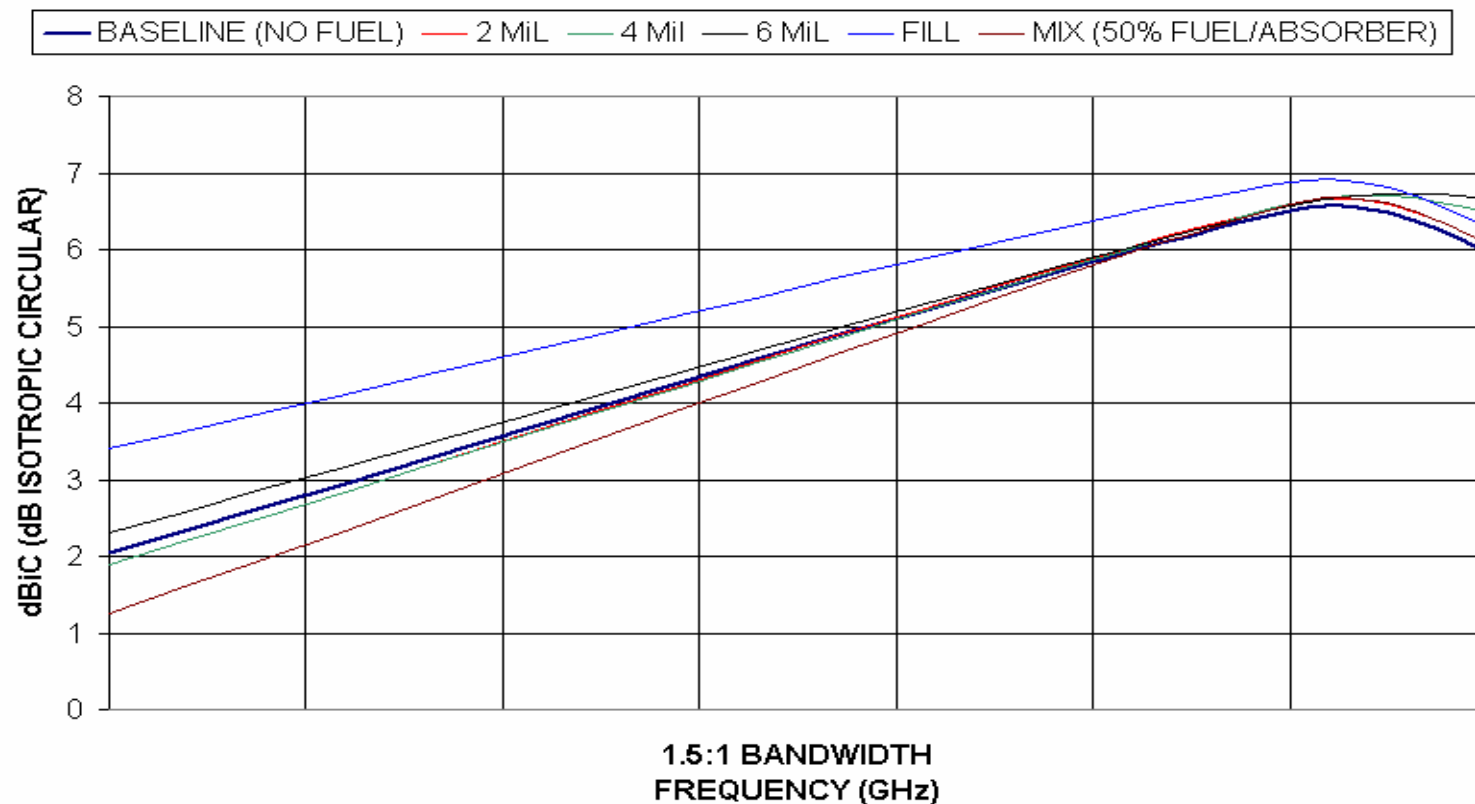


Figure 7

ANALYSIS RESULTS

KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. BEAM PEAK SQUINT (GAUSSIAN FITTED) AT MICROWAVE FREQUENCIES

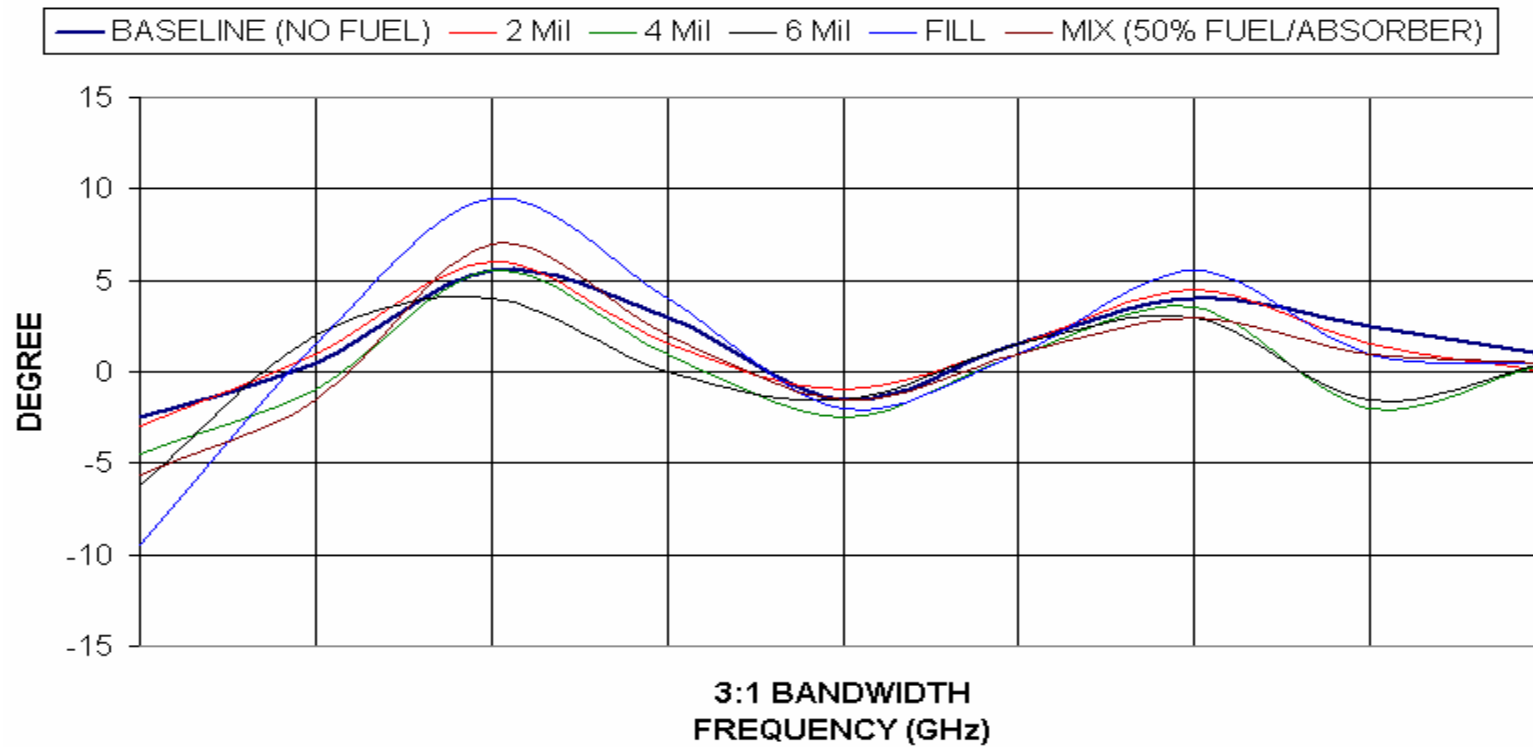


Figure 8

ANALYSIS RESULTS

**KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. BEAM
PEAK SQUINT (GAUSSIAN FITTED) AT MILLIMETER FREQUENCIES**

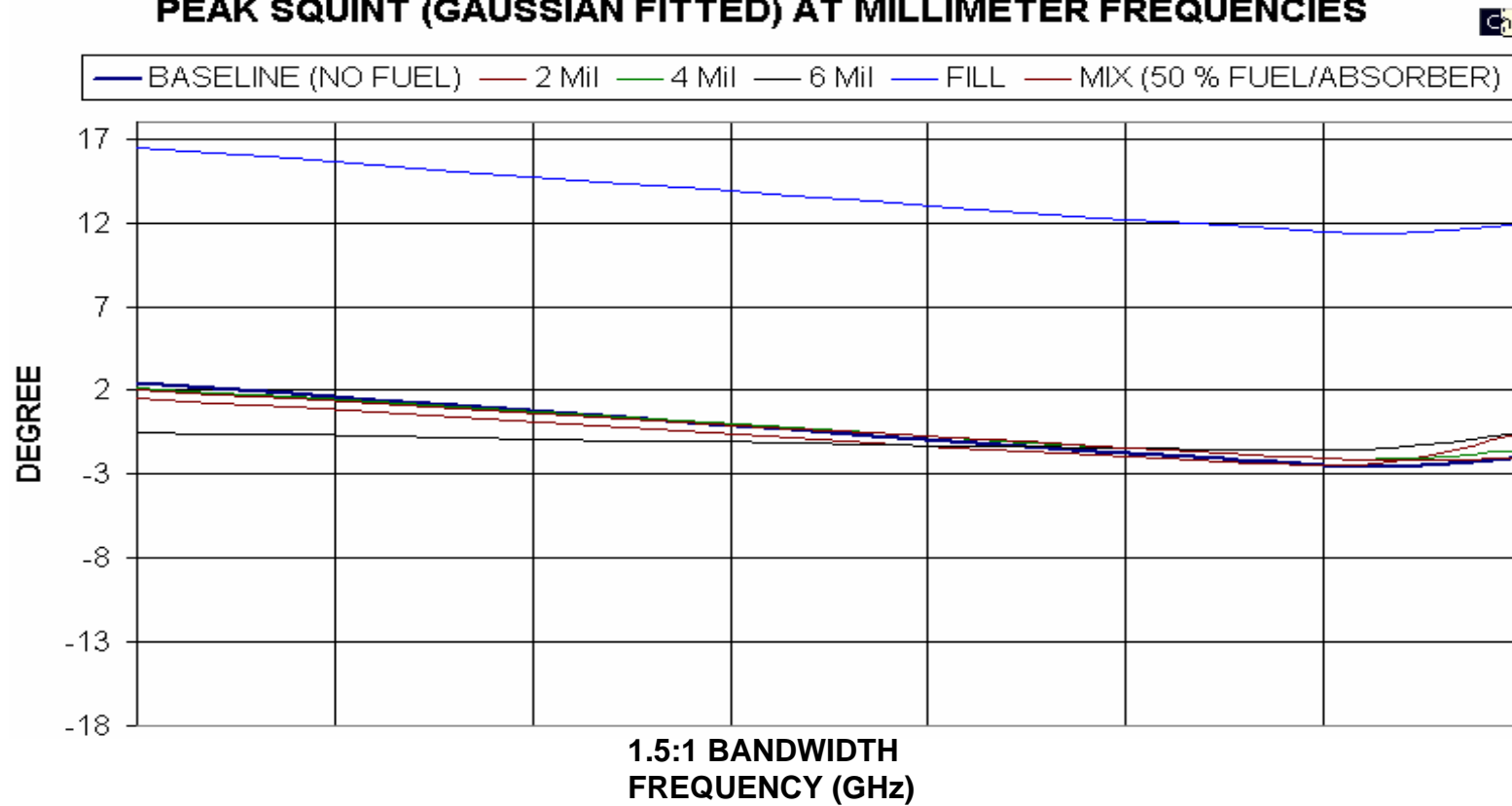


Figure 9

ANALYSIS RESULTS

KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. 10 dB BEAM WIDTH AT MICROWAVE FREQUENCIES

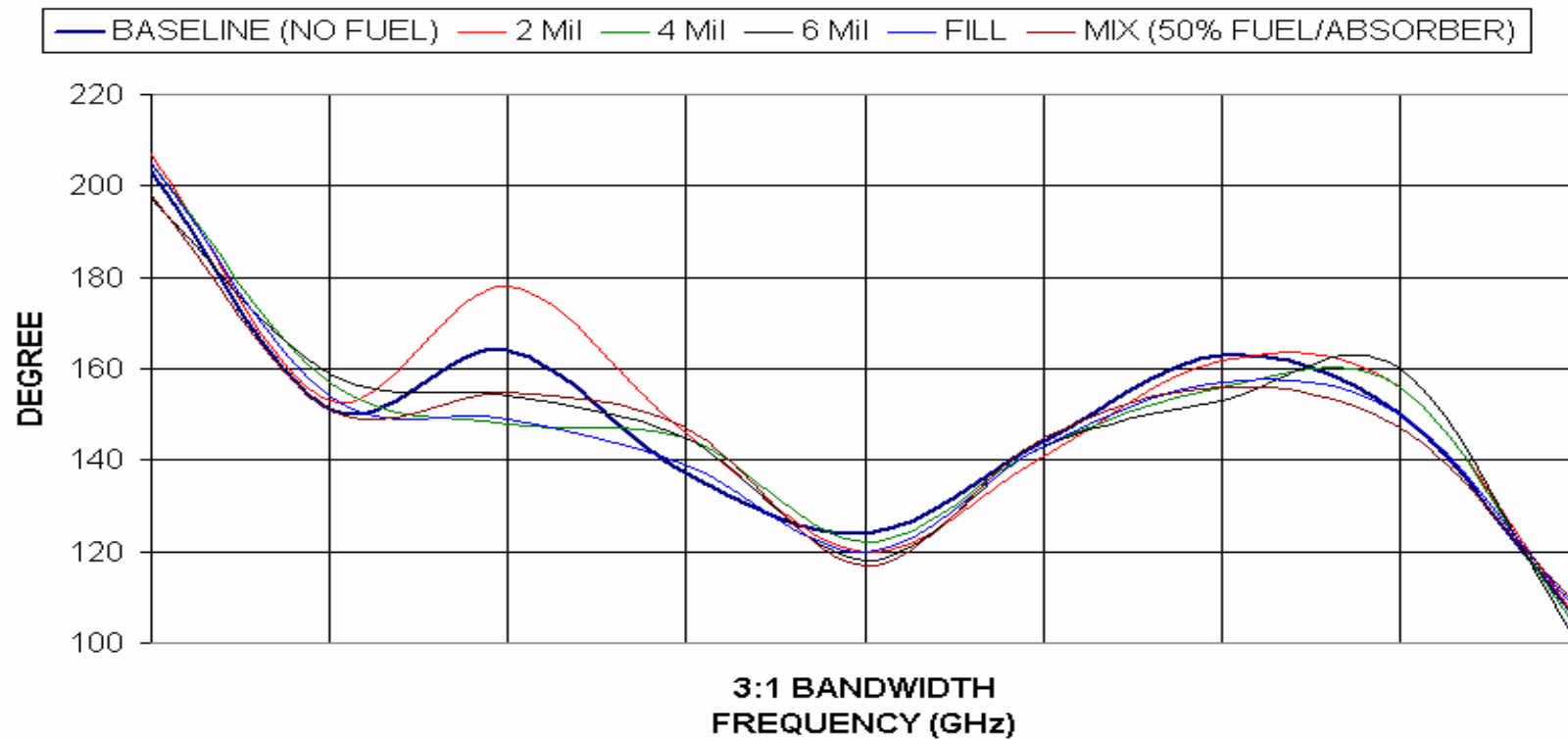


Figure 10

ANALYSIS RESULTS

KEROSENE FUEL LAYERS AND FUEL/ABSORBER MIX Vs. 10 dB BEAM WIDTH AT MILLIMETER FREQUENCIES

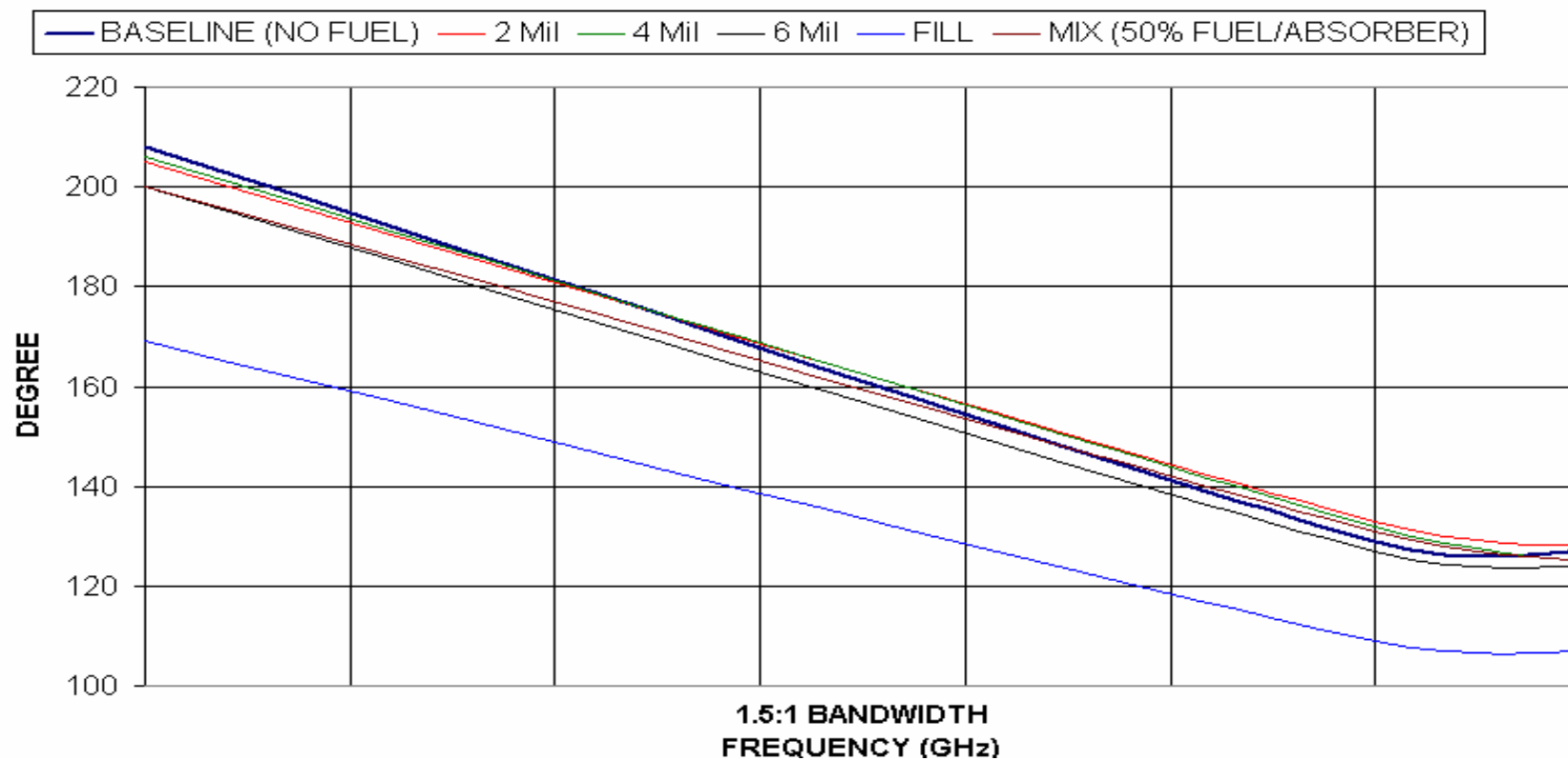


Figure 11

COMMENTS

- The enclosed presentation showed that HFSS and additional formulations can also be used to study environmental factors such as ice, and water that affect antennas performance

Questions

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